Eighth-Brick DC-DC Converter

36 - 75 V Input

12 V Output

8 A Current

Negative Logic

Description

The GAE08S12B is a new generation isolated DC-DC converter that uses an industry standard eighth-brick structure, and features high efficiency and power density, operates from an input voltage range of 36 V to 75 V, provides the rated output voltage of 12 V and the maximum output current of 8 A.

Operational Features

- Input voltage: 36 75 VOutput current: 0 8 A
- Low output ripple and noise
- Efficiency: 94.0% (12 V, 8 A)

GAE08S12B

Mechanical Features

- Industry standard eighth-brick (L x W x H):
 57.9 mm x 22.9 mm x 12.7 mm (2.28 in. x 0.9 in. x 0.50 in.)
- Weight: about 41 g

Control Features

- Remote on/off
- Remote sense
- Output voltage trim

Protection Features

- Input undervoltage protection
- Output overcurrent protection (hiccup mode)
- Output short circuit protection (hiccup mode)
- Output overvoltage protection (hiccup mode)
- Overtemperature protection (self-recovery)

Safety Features

- UL60950-1 and CSA C22.2 No. 60950-1-07
- Meet UL94V-0 flammability requirements
- RoHS6 compliant



Designation Explanation

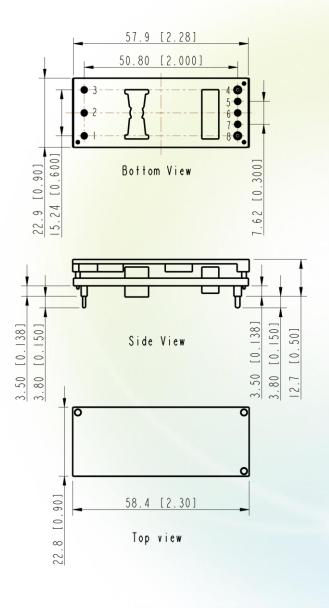
<u>GAE</u> <u>08</u> <u>S</u> <u>12</u> <u>B</u> 1 2 3 4 5

- 1 48Vin, high performance, analog control eighth-brick
- 2 Output current: 8 A
- 3 Single output
- 4 Output voltage: 12 V
- 5 With baseplate



PW11ERAZ on the label of the module is the internal model used by the manufacturer.

Mechanical Diagram



Pin Description

Pin No.	Function
1	V _{in} (+)
2	On/Off
3	V _{in} (-)
4	V _{out} (-)
5	Sense (-)
6	Trim
7	Sense (+)
8	V _{out} (+)

NOTE

- All dimensions in mm [in.]
 Tolerances: x.x ± 0.5 mm [x.xx± 0.02 in.]
 x.xx ± 0.25 mm[x.xxx ± 0.010 in.]
- 2. Pin 1-3, 5-7 are 1.00 \pm 0.05 mm [0.040 \pm 0.002 in.] diameter with 2.00 \pm 0.10 mm [0.080 \pm 0.004 in.] diameter standoff shoulders. Pin 4 and pin 8 are 1.50 \pm 0.05 mm [0.060 \pm 0.002 in.] diameter with 2.50 \pm 0.10 mm [0.098 \pm 0.004 in.] diameter standoff shoulders.



Electrical Specifications

Conditions: $T_A = 25$ °C (77°F), Airflow = 1 m/s (200 LFM), $V_{in} = 48$ V, unless otherwise notes.

Parameter	Min.	Тур.	Max.	Units	Notes & Conditions
Absolute maximum ratings					
Input voltage Continuous Transient (100 ms)	-	-	80 100	V V	-
Operating ambient temperature	-40	-	85	°C	See the thermal derating curve
Storage temperature	-55	-	125	°C	-
Operating humidity	10	-	95	% RH	Non-condensing
Input characteristics					
Operating input voltage	36	48	75	V	- ()
Maximum input current	-	-	3.2	А	V _{in} = 0 - 75 V; I _{out} = 8 A
No-load loss	-	2.4	-	W	V _{in} = 48 V; I _{out} = 0 A
Input capacitance	47	100	-	μF	Aluminum electrolytic capacitor
Inrush transient	-	-	1	A²s	-
Input reflected ripple current (peak to peak)	-	-	20	mA	Oscilloscope bandwidth: 20 MHz
Output characteristics					
Output voltage set point	11.88	12.00	12.12	V	V _{in} = 48 V; I _{out} = 8 A
Output power	0	-	96	W	-
Output line regulation	-	-	±1	%	V _{in} = 36 - 75 V; I _{out} = 8 A
Output load regulation	-		±1	%	V _{in} = 48 V; I _{out} = 0 - 8 A
Regulated voltage precision	-		±3	%	V _{in} = 36 - 75 V; I _{out} = 0 - 8 A
Temperature coefficient	-	-	±0.02	%/°C	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C } (-40^{\circ}\text{F to } +185^{\circ}\text{F })$
External capacitance	220	470	5000	μF	Aluminum electrolytic capacitor
Output current	0	-	8	Α	-
Output ripple and noise (peak to peak)	-	50	200	mV	Oscilloscope bandwidth: 20 MHz
Output voltage Trim range	80	-	110	%	-
Output voltage overshoot	-	-	5	%	The whole range of V_{in} , I_{out} and T_A
Output voltage delay time	-	-	200	ms	from V _{in} connection to 10% V _{out}
Output voltage rise time		12	20	ms	from 10%V _{out} to 90%V _{out}
Switching frequency	-	260	-	kHz	-

Electrical Specifications

Conditions: $T_A = 25$ °C (77°F), Airflow = 1 m/s (200 LFM), $V_{in} = 48$ V, unless otherwise notes.

Parameter	Min.	Тур.	Max.	Units	Notes & Conditions
Protection characteristics					
Input undervoltage protection Startup threshold Shutdown threshold Hysteresis	31.0 30.0 1.0	34.5 32.5	36.0 35.0 3.0	V V	-
Output overcurrent protection	8.8	-	14.0	Α	Hiccup mode
Output short circuit protection	-	-	-	-	Hiccup mode
Output overvoltage protection	14.4	-	18.0	V	Hiccup mode
Overtemperature protection Threshold Hysteresis	100 5	-	125 -	°C °C	Self-recovery The values are obtained by measuring the temperature of the PCB near the thermal resistor.
Dynamic characteristics					
Overshoot amplitude Recovery time		-	600 400	mV μs	Current change rate: 0.1 A/µs load : 25% - 50% - 25%; 50% - 75% - 50%
Overshoot amplitude Recovery time	-	-	800 500	mV μs	Current change rate: 1 A/µs load : 25% - 50% - 25%; 50% - 75% - 50%
Efficiency					
100% load	92.0	94.0	-	%	V _{in} = 48 V; I _{out} = 8 A
50% load	91.0	93.0	-	%	V _{in} = 48 V; I _{out} = 4 A
Isolation characteristics					
Input-to-output Isolation voltage	-	-	1500	V DC	Basic Isolation
Other characteristics					
Remote on/off voltage Low level High level	-0.7 3.5	-	1.2 12	V V	-
On/Off current Low level High level	1 1	-	1.0	mΑ μΑ	-
Reliability characteristics					
Mean time between failures (MTBF)	-	2.5	-	Million hours	Telcordia SR332; 80% load; Airflow = 1.5m/s (300LFM); T _A = 40°C (104°F)

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Characteristic Curves

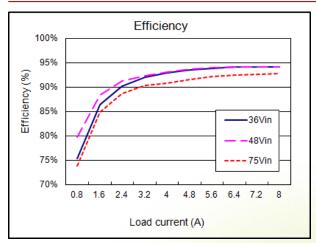


Figure 1: Efficiency $(T_A = 25^{\circ}C \text{ or } 77^{\circ}F)$

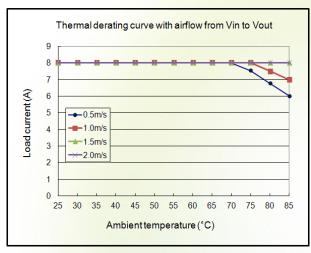


Figure 3: Thermal derating with airflow from V_{in} to V_{out} (V_{in} = 48 V; V_{out} = 12 V)

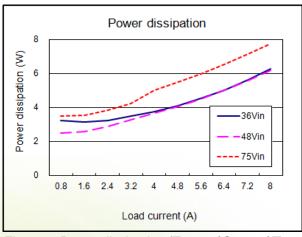


Figure 2: Power dissipation ($T_A = 25^{\circ}C$ or $77^{\circ}F$)

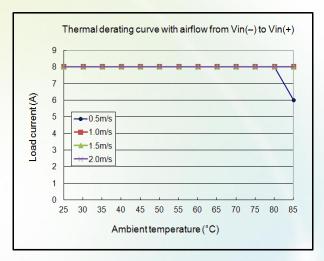


Figure 4: Thermal derating with airflow from $V_{in}(-)$ to $V_{in}(+)$ ($V_{in} = 48 \text{ V}$; $V_{out} = 12 \text{ V}$)

Typical Waveforms

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- 1. During the test of input reflected ripple current, the input terminal must be connected to a 12 μH inductor and a 220 μF electrolytic capacitor.
- 2. Point B, which is for testing the output voltage ripple, is 25 mm (0.98 in.) away from the Vout(+) pin.

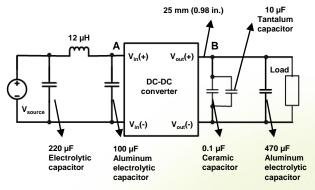


Figure 5: Test set-up diagram

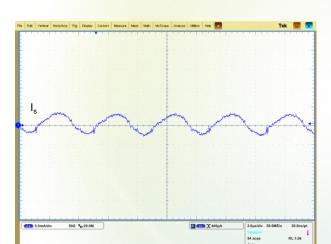


Figure 7: Input reflected ripple current (for point A in the test set-up diagram, $V_{in} = 48 \text{ V}$, $V_{out} = 12 \text{ V}$, $I_{out} = 8 \text{ A}$)

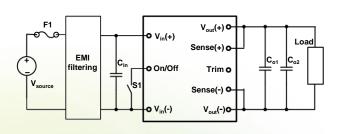


Figure 6: Typical circuit applications

F1: 5 A fuse (fast blowing)

C_{in}: The high-frequency, low equivalent series resistance (ESR) electrolytic capacitor (100 μF/100 V) is recommended.

 $C_{o1}.$ The 1 μF ceramic capacitor is recommended. $C_{o2}.$ The 470 μF aluminum electrolytic capacitor is recommended.

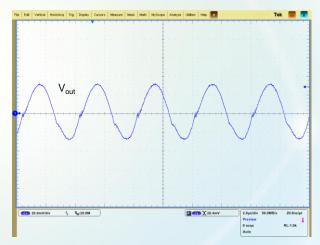


Figure 8: Output voltage ripple (for point B in the test set-up diagram, $V_{in} = 48 \text{ V}$, $V_{out} = 12 \text{ V}$, $I_{out} = 8 \text{ A}$)

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Typical Waveforms

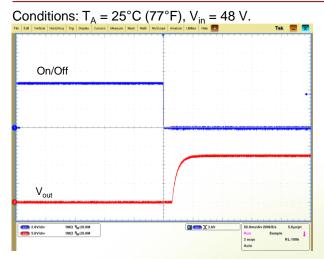


Figure 9: Startup from On/Off

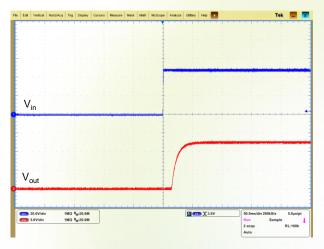


Figure 11: Startup by power on

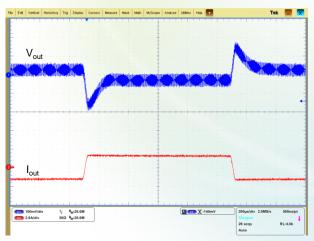


Figure 13: Output voltage dynamic response (Load: 25% - 50% - 25%, di/dt = 0.1 A/µs)

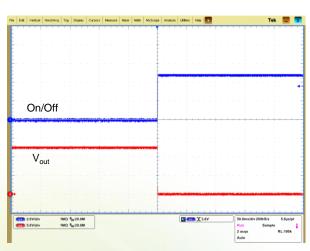


Figure 10: Shutdown from On/Off

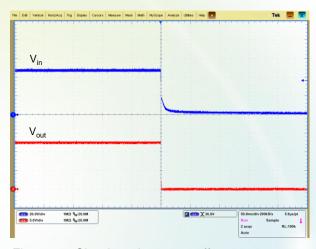


Figure 12: Shutdown by power off

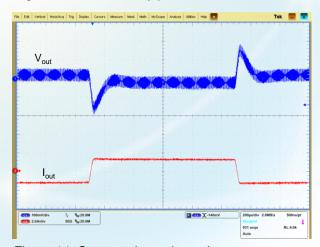


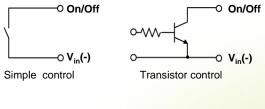
Figure 14: Output voltage dynamic response (Load: 50% - 75% - 50%, di/dt = 0.1 A/µs)



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Remote On/Off

Logic Enable	On/Off Pin Level	Status
Negative	Low level	On
logic	High level or left open	Off



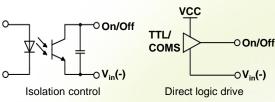


Figure 15: various circuits for driving the On/Off pin

Remote Sense

This function is used to compensate for voltage drops on R_w. The Sense(+), Sense(-), V_{out}(+), and V_{out}(-) terminals should meet the following requirements:

 $[V_{out}(+) - V_{out}(-)]$ - $[Sense(+) - Sense(-)] \le 10\% \times V_{nom}$ (V_{nom} is the rated output voltage.)

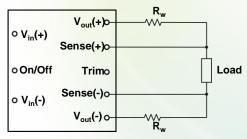


Figure 16: Configuration diagram for remote sense

R_w indicates the line impedance between the output terminal and the load

If the remote sense function is disabled, the Sense(+) terminal directly connects to the $V_{out}(+)$ terminal and the Sense(-) terminal directly connects to the $V_{out}(-)$ terminal.

Output Voltage Trim

The output voltage can be adjusted according to the trim range specification by using the Trim pin.

Trim Up

The output voltage can be increased by installing an external resistor between the Trim pin and the Sense(+) pin.

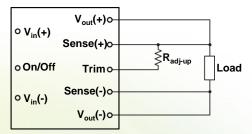


Figure 17: Configuration diagram for Trim up

The relationship between R_{adj-up} and V_{out}:

$$R_{adj-up} = \frac{5.1 \times V_{nom} \times (100 + \Delta)}{1.225 \times \Lambda} - \frac{510}{\Lambda} - 10.2(k\Omega)$$

$$\Delta = \frac{V_{out} - V_{nom}}{V_{nom}} \times 100$$

NOTE

- I. If the Trim pin is not used, it should be left open.
- Ensure that the actual output power does not exceed the maximum output power when raising the voltage.

Trim Down

The output voltage can be decreased by installing an external resistor between the Trim pin and the Sense(-) pin.

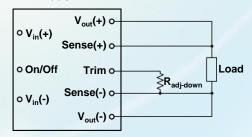


Figure 18: Configuration diagram for Trim down

The relationship between R_{adj-down} and V_{out}:

$$R_{adj-down} = \frac{510}{\Delta} - 10.2(k\Omega) \qquad \Delta = \frac{V_{nom} - V_{out}}{V_{nom}} \times 100$$

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Input Undervoltage Protection

The converter will shut down after the input voltage drops below the undervoltage protection threshold for shutdown. The converter will start to work again after the input voltage reaches the input undervoltage protection threshold for startup. For the Hysteresis, see the Protection characteristics.

Output Overcurrent Protection

The converter equipped with current limiting circuitry can provide protection from an output overload or short circuit condition. If the output current exceeds the output overcurrent protection threshold, the converter enters hiccup mode. When the fault condition is removed, the converter will automatically restart.

Output Overvoltage Protection

When the voltage directly across the output pins exceeds the output overvoltage protection threshold, the converter will enter hiccup mode. When the fault condition is removed, the converter will automatically restart.

Overtemperature Protection

A temperature sensor on the converter senses the average temperature of the module. It protects the converter from being damaged at high temperatures. When the temperature exceeds the overtemperature protection threshold, the output will shut down. It will allow the converter to turn on again when the temperature of the sensed location falls by the value of Overtemperature Protection Hysteresis.

MTBF

The MTBF is calculated according to the Telcordia, SR332 Method 1 Case3.

Recommend Reverse Polarity Protection Circuit

Reverse polarity protection is recommended under installation and cabling conditions where reverse polarity across the input may occur.

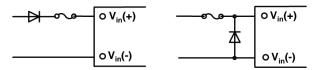


Figure 19: Recommend reverse polarity protection circuits

Recommended Fuse

The converter has no internal fuse. To meet safety and regulatory requirements, a 5 A fuse is recommended.



The fuse current should be 1.5 to 2 times the maximum operating current in actual use.

EMC

For the acceptance standard, see the *DC-DC* Converter EMC Acceptance Manual.

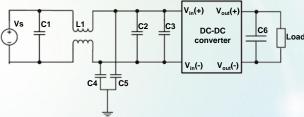


Figure 20: EMC test set-up diagram

C1: Surface mount device (SMD) ceramic capacitor (100 V/1000 nF/X7R/1210)

C2: SMD ceramic capacitor (100 V/100 nF/±10%/X7R/1206)

L1: Common-mode inductor (single phase, 1320 μ H/ \pm 25%/4 A/R5K/ 21 mm x 21 mm x 12.5 mm [0.83 in. x 0.83 in. x 0.49 in.]). The chip component with the same specifications can also be used.

C4,C5: High-pressure resistant chip ceramic capacitor (22 nF/1000 V/X7R/1210)

C3: Aluminum electrolytic capacitor (100 µF)

C6: Aluminum electrolytic capacitor (470 µF)

Qualification Testing

Parameter	Units	Condition
High Accelerated Life Test (HALT)	4	Lowest operating temperature: -60°C (-76°F); highest operating temperature: 120°C (248°F); vibration limit: 40 G
Thermal Shock	32	500 temperature cycles between -40°C (-40°F) and +125°C (+257°F) with the temperature change rate of 20°C (68°F) per minute; Lasting for 30 minutes both at -40°C (-40°F) and +125°C (+257°F)
Temperature Humidity Bias (THB)	16	Maximum input voltage; 85°C (185°F); 85% RH; 1000 operating hours under lowest load power
High Temperature Operation Bias (HTOB)	16	Rating input voltage; air flow:0.5 m/s (100 FLM) to 5 m/s (1000 FLM); 1000 operating hours under 50% - 80% load power; air temperature:45°C(113°F) - 55°C(131°F)

Thermal Consideration

Thermal Test Point

Sufficient airflow should be provided to ensure reliable operating of the converter. Therefore, thermal components are mounted on the top surface of the converter to dissipate heat to the surrounding environment by conduction, convection and radiation. Proper airflow can be verified by measuring the temperature at the middle of the base plate.

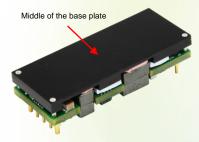


Figure 21: Thermal test point



The temperature at the thermal test point on the converter cannot exceed 100°C (212°F). Otherwise, the converter will be protected against overtemperature and will not operate properly.

Power Dissipation

The converter power dissipation is calculated based on efficiency. The following formula reflects the relationship between the consumed power (P_d), efficiency (η), and output power (P_o): $Pd=Po(1-\eta)/\eta$

Mechanical Consideration

Installation

Although the converter can be mounted in any direction, free airflow must be taken.

Soldering

The converter is compatible with standard wave soldering techniques. For wave soldering, the converter pins should be preheated for 20 to 30 seconds at 110°C (230°F), and wave soldered at 260°C (500°F) for less than 7 seconds.

For hand soldering, the iron temperature should be maintained at 350°C (662°F) to 420°C (788°F) and applied to the converter pins for less than 10 seconds.

The converter can be rinsed using the isopropyl alcohol (IPA) solvent or other proper solvents.

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